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## Performance Improvement of a Slotted Solid Rotor Induction Motor with High Temperature Superconder Coating

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#### PAPER INFO

ABSTRACT

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Keywords: Bi-2223/Ag Super Conductor Critical Current Slotted Solid Rotor Starting Torque This paper analyzes an induction motor with high temperature superconducter (HTS) coated slotted solid rotor. By slotting the solid rotor, the electromagnetic torque near synchronous speed will increase but the starting torque will decrease. To improve starting torque, rotor slots are coated with HTS materials. Using HTS material vary the rotor resistance to great extends in starting step and this means the starting torque will be maximized. Also the torque near synchronous speed will be higher than before because of the resistance of the HTS is completely zero in this period. This would help the torque become more than the case without any coating. It is concluded that, HTS coating of slotted solid rotor will increase the starting torque by 42 and 74% than smooth type rotor and slotted one, respectively. For case-study motor, the torque near synchronous speed these improvements were 75% and 33%, respectively. The performance of the proposed motor is simulated using finite element method (FEM).

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## **1. INTRODUCTION**

Nowadays, induction motors are known as an extremely popular consumer in all industrial applications. These motors are cost effective and easy to maintain. Furthermore, self-starting is considered as a great advantage of these type. The resistance of rotor plays a vital role in the performance of squirrel cage induction motors including, starting torque and efficiency. A small amount of rotor resistance results in, a high efficiency at full load operation and a low amount of starting torque. On the other hand, a high starting torque and low efficiency at full load operation will be obtained as a result of a large rotor resistance [1-3]. In order to have a variable resistance in the rotor circuit, an external resistance is always needed. But, this issue is only applicable for the wound rotor induction motors. Thus, the major problem refers to immutability of resistance in the rotor of conventional squirrel cage motors which is made out of copper or aluminum.

It has been observed that, applying the solid rotor

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instead of squirrel cage one is a great alternative in order to obtain the high load torque and improve the starting torque. Furthermore, the mentioned rotor has high mechanical robustness and manufacturing of that is cost effective. However, the use of solid rotor induction motor is restricted to specific applications due to two significant drawbacks including, low power density and low electrical conductivity compared to other alternatives [4].

So far, several investigations have been conducted with concentration on fabrication and characteristics of induction motors, using HTS materials. Investigation conducted by Sim et al. [5], used HTS wire instead of conventional end rings and also the bars of a specified induction motor. BSCCO-2223, is the type of HTS material with critical current of 115 A. Two different tests including the block rotor and also the load test were carried out in case and the results showed that, after 63% of the full load the resistance starts to emerge in the HTS tapes at continuous behavior. It should be noted that, before this point, the rotating speed of the HTS motor is the same as the synchronous speed. An induction motor is fabricated, using Bi-2223/Ag HTS material in end rings and also the bars [6]. The test was

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conducted for various input voltages and the electrical equivalent circuit was used for analyzing the performances of the motor. The better starting and also accelerating torque of the HTS motor compared with the conventional one is resulted in as a great advantage of newly fabricated motor. Contribution by Song et al. [7], discussed about an experimental model of an induction motor which is immersed in liquid nitrogen bath with a superconducting secondary conductors, using Bi-2223/Ag HTS material. The mentioned test showed different results including; decrement in the resistance of the superconducting rotor with decreasing the slip and also achieving a better starting torque. Sekiguchi et al. clarified the application of [8], HTS Induction/Synchronous machines (HTS-ISM) in electrical vehicles. A 20 KW class of HTS-ISM is designed and also fabricated with DI-BSCCO superconductor. In accordance to the results, it is concluded that, the efficiency has greatly improved in the mentioned machine. But unfortunately, the steady state slip rotation was not considered. It should be noted that, the main contribution of this study is achieving a fully superconducting machine including an excellent synchronous rotation, containing a higher speed, for the first time in the world. Nakamura e al. [9] studied about design and also fabrication of the second generation of 20 KW class of HTS-ISM, which is used in the transportation equipment. The volume of the second generation HTS-ISM is reduced about 70% compared to the first class one. But the major principle of this paper is, the great enhancement of the torque and also the power density, simultaneously. BSCCO-2223, is applied as the HTS material in the rotor windings. It is noteworthy that, the system is cooled in the liquid nitrogen.

It is noted, some other investigations have focused on slotted solid rotor motor for the improvement of motor performance. As a result, Aho et al. [10] mentioned the higher are the saturation of magnetic flux density and electrical conductivity of the motor materials, the higher is the torque of motor. Therefore, the saturation of flux density and the electrical conductivity of the materials have direct relation with the torque of motor. Sharma et al. [11] aimed at maximizing the electromagnetic torque through depositing layers of Cu plus Fe-Co on the rotor surface. Other investigations indicated that, a better output torque and also a higher efficiency were considered as two significant advantages of a slotted solid rotor induction motor compared with a smooth one [12–14].

This study proposed deposition of a thin layer of HTS on the outer face of the solid rotor slots in order to improve the performance of the induction motor. Consequently, by means of the mentioned layer of HTS, the rotor resistance can have a great amount of incremental variation [15]. During the starting period of

induction motor, the slip of the motor is equal to one. As a result, two significant issues simultaneously occur. The first issue includes enormous induced current in the rotor core and the second one is the increased frequency of the rotor. The two important mentioned issues make HTS layer quench during the starting period. Consequently, a high starting torque can be achieved. But during the normal starting condition and after raising the speed and the torque of the motor, the HTS layer acts as a superconductor due to decrement of current and frequency of the rotor. After the perfect recovery of the HTS from quenching, the resistance of the solid rotor gets quite small due to zero resistance of the layer. In such situations, power loss in rotor circuit is minimum and only depends on some other factors including the rotor core and etc. As a result, a large current can be induced in the rotor circuit of the proposed motor at low levels of slip which leads to a high torque near the synchronous speed. It should be noted that, for a widespread range of load, the slip is equal to zero. Ultimately a real case-study motor with the proposed idea is simulated using finite element method. The obtained results confirmed the theoretical concepts.

## 2. PROPOSED SCHEME PRINCIPLE

The starting torque is always known as an important principle in the performance of induction motors. By the same token, solid rotor induction motors are widely used, because of their inherent characteristics, for improvement of the starting torque. On the other hand, a higher torque near the synchronous speed is always desirable. To this end, a slotted solid rotor is proposed. In fact, this type of rotor increases the penetration of the flux lines in the rotor core by means of the embedded slots. Thus, the existing volume for energy conversion will enhance compared to smooth solid rotor and consequently a high output torque in low slips is obtained [14, 16]. In other words, the electromagnetic torque in a solid rotor induction motor is generated due to interaction between the stator current and the eddy current, induced in the rotor. Therefore, in order to calculate the electromagnetic torque, the eddy current loss should be calculated at first. The mentioned loss is obtained by the following equation [17]:

$$P_e = \sigma^{-1} \iint |J|^2 \, dx \, dy$$
  
Induced region (1)

where  $\sigma$  and J are the conductivity coefficient of the rotor and induced current density, respectively. By considering  $\omega$  as the rotor angular velocity, the electromagnetic torque can be obtaind as follows:

$$\tau = \frac{P_e}{w} \tag{2}$$

It is concluded from Equation (1) that, the enhancement of the space at which the current is induced, causes the increment of the electromagnetic torque. It is worth noting that, a slotted rotor increases the flux density in the rotor  $B_r$ . The electromagnetic torque  $\tau$ , can be maximized in low slips in accordance to the following equation.

$$\tau = kB_s \times B_r \tag{3}$$

where,  $B_s$  is the stator flux density and k is a constant which is defined as follows:

$$k = -\frac{P}{2} \frac{gDL\pi}{2\mu_0} \sin\alpha \tag{4}$$

where  $\mu_0$ , L, D, g and P, represent absolute permeability, axial length of the air gap, the diameter of the air gap, the length of the air gap and the number of poles, respectively. It should be noticed that,  $\alpha$  is the angle between the magnetic axis of the rotor and stator. As previously mentioned, penetration of the flux lines is increased in the rotor core by means of the embedded slots. Figure 1 shows this flux penetration in the slotted rotor and B-H curve of rotor material.

As previously mentioned, the starting torque is a significant principle which should always be considered, in performance of induction motors for various applications. Accordingly, applying the slotted rotor causes reduction of rotor electric resistance and also flowing a large amount of induced current in the rotor core which can eventually lead to a lower starting torque in slotted rotors compared to smooth types. In order to produce a high starting torque, a large starting resistance is required. On the other hand, during the normal accelerating condition the key point is minimization of power loss. To this end, a lower rotor resistance is always required during the normal accelerating period. As a result, a high electromagnetic torque is obtained in low levels of slip. The mentioned change in resistance, at the starting and also during normal operation of the motor leads to the best performance of an induction motor. In order to achieve the mentioned desirable performance, a thin layer of HTS is used on the slots surface. Hereby, the rotor resistance can vary as desired. As discussed before, the electric resistance should have its maximum value, at the start time. This issue is easily justified by the high current and frequency, applied to the rotor at this time. It should be noted that, in such circumstances, the layer of HTS is in quenching mode. But during the normal running period, the situation is different. In fact, the layer of HTS performs as a perfect conductor due to low current and also low frequency. Therefore, the resistance of the HTS will be equal to zero and the rotor resistance will be quite small. Consequently, the induction torque increases at low amounts of slip and a higher efficiency is obtained.



Figure 1. a) Flux penetration in the slotted rotor core; b) B-H curve of rotor material

To illustrate the above points, Figure 2 compares the torque-speed characteristics of a 4-pole, 50-Hz smooth solid-rotor and a slotted solid-rotor and also a HTS coated slotted solid-rotor induction motor. It can obviously be seen that, the HTS coated slotted motor produces a higher starting torque compared to other motors and the mentioned torque is really near the synchronous speed unlike others. As another amazing reality extracted from Figure 2, it could be said that, the synchronous torque, for a HTS coated slotted rotor, occurs at about 5 Nm in the synchronous speed.



**Figure 2.** Comparison of torque-speed characteristics of a 4pole, 50-Hz smooth solid-rotor with the slotted solid-rotor and HTS coated slotted solid- rotor induction motor

## **3. FINITE ELEMENT ANALYSIS**

A 2-D finite-element package is used to simulate the solid rotor induction motor. A time stepping finite element method is also needed in order to solve the magnetic field because of the non-linearity of rotor material and also the movement of the rotor. The calculation is based on solution of the magnetic vector potential A [4].

The curl of magnetic vector potential is equal to the magnetic flux density **B**, as in the following equation:

$$\nabla \times A = B \tag{5}$$

On the other hand, the divergence of the magnetic vector potential is equal to zero as follows:

$$\nabla \cdot A = 0 \tag{6}$$

The following equation can be obtained by substituting the magnetic vector potential definition in the induction law:

$$\nabla \times E = -\nabla \times \frac{\partial A}{\partial t} \tag{7}$$

As a result, the electric field equation can be written as follows:

$$E = -\frac{\partial A}{\partial t} - \nabla \phi \tag{8}$$

where, E and  $\phi$  are the electric field strength and the electric scalar potential, respectively.

The eddy current density in the conducting zone can be stated as follows:

$$J_e = \sigma E = -\sigma \frac{\partial A}{\partial t} - \sigma \nabla \phi \tag{9}$$

where,  $\sigma$  represents the electric conductivity and *E* was already defined in Equation (8). The governing equation can be written in accordance to Ampere's law and magnetic vector potential as follows:

$$\nabla \times (\frac{1}{\mu} \nabla \times A) = J_e \tag{10}$$

where,  $\mu$  is the magnetic permeability.

It is worth mentioning that, the induction torque is produced by an interaction between stator current of voltage source and the eddy current induced, in rotor. By assuming H and V as the magnetic field intensity and the volume, respectively. The related equation to the output torque  $T_{out}$  with angular displacement  $\theta$ , can be displayed as follows:

$$T_{out} = \frac{\partial}{\partial \theta} \int_{V} \int_{0}^{H} B dH dV$$
(11)  
$$\nabla \cdot A = 0$$
  
$$\nabla \times E = -\nabla \times \frac{\partial A}{\partial t}$$

**3.1. HTS Material Modeling** 

The nonlinear

characteristics of the HTS material can be expressed as follows [18]:

$$E = E_c \left(\frac{\left|\vec{J}\right|}{J_c}\right)^n \tag{12}$$

where  $\vec{E}$  is the electrical field and  $\vec{J}$  represent the current density vectors of the HTS. Furthermore,  $E_c$ ,  $J_c$  and *n* are the critical electrical field, critical current density and exponential factor of the HTS material, respectively.

The Kim–Anderson model, is known as a significant model in order to introduce the HTS resistivity  $\rho$ , which is expressed as follows [19]:

$$\rho(E,B) = \frac{E_c^{\frac{1}{n(B)}}}{J_c(B)} \left| \overline{E} \right|^{\frac{n(B)-1}{n(B)}} + \rho_0$$
(13)

where  $\rho_0$  is the additional resistivity, and variables which are dependent to the magnetic flux density *B*, are evaluated by the following equations:

$$n(B) = \frac{n}{1 + \frac{B}{B_0}}$$
(14)

$$J_{c}(B) = \frac{J_{c0}}{1 + B_{B0}}$$
(15)

where  $J_{c0}$ ,  $B_0$  and  $n_0$  are the reference values of the critical current density, the magnetic flux and the exponent factor, respectively [20].

#### 4. CASE STUDY SPECIFICATIONS

**4. 1. Specifications of the Studied Motor** In this study, a 1.05 kW solid rotor induction motor which was previously designed and fabricated by Sharma et al. [11] which is taken as a real case study. In the mentioned motor, the stator and complete scale are the same. But the difference is that, the slots of the solid rotor are coated with HTS material. The input voltage has a value of 250 volts and the starting and the normal current values are 15 and 3.7 A, respectively. The complete data of the motor is described in Table 1. It should be noted that, the fabricated axially slotted solid rotor reported in literature [16] as shown in Figure 3.

**4. 2. Specifications of the HTS Material** Figure 4 shows a cross section shape, related to the placement of the HTS material layer in the rotor slots. It should be noticed that, the magnetic flux density of the studied motor, both in starting and steady state condition, considering the HTS material in rotor slots are also

<b>TABLE 1.</b> Data of the case study motor				
Stator	Outer diameter	176 mm		
	Inner diameter	115 mm		
	Stack length	46.5 mm		
	Number of slots	48		
	Conductor per slot	70		
	Slot depth	11.5 mm		
Rotor	Outer diameter	114 mm		
	Number of slots	12		
	Inertia (J <sub>r</sub> )	0.64 <sup>2</sup>		



(a)



**Figure 3.** a) Fabricated slotted solid rotor b) Exprimental Setup [16]



Figure 4. Part of cross section of the case study motor with HTS layer on the rotor slot surface

shown in Figures 5(a) and 5(b), respectively. In these figures the steady state operation regards to synchronous speed situation.

The operation of all types of HTS materials are restricted within three significant parameters including; critical temperature, current density and magnetic field [21], as shown in Figure 6.



**Figure 5.** Magnetic flux distribution of the new motor at a) Starting condition; b) Steady state operation



Figure 6. Superconductivity in a 3D space defined by current density, temperature and magnetic field [20]

It should be noted that, the HTS material can be reverted to its normal condition (no superconductivity exhibition), if any of the three mentioned restrictions is exceeded. Therefore, the complete specifications of the applied HTS material layer which are given in Table 2, [21, 22] seem to be essential. It should be noted that, the operating temperature of the specified HTS material is at 77K which is known as the liquid nitrogen temperature.

In accordance to the mentioned Table, the critical magnetic field is the only important parameter which is not given in the operating temperature. Thus a relation for clarification of this significant parameter at any ambient temperature is needed.

Material	Bi-2223/Ag (DI- BSCCO-Type H)
Width	4.3±0.3mm
Thickness	0.22±0.2 mm
Max tensile stress @ RT	100 MPa
Max tensile stress @ 77 K	130 MPa
Critical current @ 77 K and self-field	200 A
Critical current density @ 77 K and self-field	$200 \text{ A/mm}^2$
Critical current @ 4.2 k and 17 T	400 A
Critical current density @ 4.2 k and 17 T	$400 \text{ A/mm}^2$
Max bending diameter @ RT	70 mm
critical magnetic field at 0k	184 T
Critical temperature	110 K

The superconductors always exclude the magnetic fields, until their critical magnetic field is exceeded by the specified magnetic field. The critical field is defined at 0k and is equal to zero at critical temperature. Thus, the enhancement of the temperature leads to decrement of the critical magnetic field. The relation of the critical magnetic field at any arbitrary temperature below the critical temperature can be expressed as follows [23]:

$$B_c \approx B_c \left(0\right) \left[ 1 - \left(\frac{T}{T_c}\right)^2 \right]$$
(16)

where  $B_c(0)$  and  $T_c$  are the critical magnetic field at 0k and critical temperature of the applied superconductor, respectively; which were already given in Table 2. Parameter T, represents the arbitrary temperature at which, the critical magnetic field is obtained.

4. 3. Critical Current Determination The critical current of the HTS layer is always considered as a determinative factor for a proper resistance adjustment of the slotted solid rotor in both starting and normal running condition of the motor. The measured critical current in Table 2, is obtained by using a standard four point configuration at liquid nitrogen [24]. It should be noted that, the voltage criterion is approximated to be 1  $\mu$ V/cm.

The external perpendicular magnetic field dependence of the critical current has been evaluated at different temperatures, ranging from 4.2 K to 90 K [25], as shown in Figure 7.

As a result, a large amount of critical current prevents the quenching of HTS material during the starting time. On the other side, by considering a small amount of critical current, the recovery of the HTS material from quenching will never occur in normal running operation. Critical current of HTS layer was 200 A at 77 k and self-field.



Figure 7. Critical current in presence of external field at different temperatures [24]

## **5. SIMULATION ESULTS AND DISCUSSIONS**

All the simulation was done by the simulation software Ansoft Maxwell 10. Figure 8 has depicted the speed versus time charecterstic in order to illustrate the capabilty of the HTS coated induction motor. As it can be clearly seen in Figure 8, the induction motor can amazingly reach to synchronous speed under the load torque of 5 Nm and also rotor moment of inertia, Jr, whereas, the un-coated slotted motor works with approximate slip of 3% in the same situation. As previously mentioned, in such circumstances the motor can easily reach to synchronous speed because of two important reasons including low resistance of HTS deposited rotor near the synchronous speed and also the existence of the electromagnetic torque in this speed.

Figure 9 also shows that under load torque of 5 Nm and five times of rotor inertia, the proposed motor reaches to synchronous speed after longer period of time compared to the previous situation. In this term, the slotted solid rotor runs at the slip of 6%.

Finally, in order to achieve a better approval for the capability of the new proposed method, another simulation is conducted. The results of the mentioned simulation, shown in Figure 10 demonstrate that, for



Figure 8. Comparison between speed-time responses of uncoated slotted solid and HTS coated slotted rotor motor for 5 Nm load torque and rotor moment of inertia



**Figure 9.** Comparison between speed-time responses of uncoated slotted solid and HTS coated slotted rotor motor for 5 Nm load torque and 5\*Jr



Figure 10. Comparison between speed-time responses of uncoated slotted solid and HTS coated slotted rotor motor for 6 Nm load torque and rotor moment of inertia

inertia of the rotor and under constant 6 Nm load torque, HTS slotted solid-rotor motor works at low slip of 2.3 % whereas, the usual slotted one will oscillate with a high slip of 16%.

The approximate performance improvement values of different rotor structures with various load torques and moment of inertia are given in Table 3.

As a result of the simulations, it is concluded that, coating the slotted solid rotor with HTS materials causes

**TABLE 3.** Performance improvement of induction motor with different types of rotor

Rotor type	Load torque (Nm)	Moment of inertia	Slip (%)	Time to reach the synchronous speed (s)
	5	1	3	0.2 s to 1450 rpm
Slotted rotor	5	5	6	0.7 s to 1450 rpm
	6	1	16	0.5 s to 1250 rpm
	5	1	0	0.4 s to 1500 rpm
HTS coated slotted rotor	5	5	0	2.5 s to 1500 rpm
	6	1	2.3	0.2 s to 1500 rpm

an increment in starting torque, which is approximated to be 42 and 74% than smooth type rotor and slotted one respectively. For the case-study motor, the torque near synchronous speed these improvement were 75 and 33% respectively.

## **6. CONCLUSION**

In this paper, finite element method was implemented to analyze a new structure of induction motor with HTScoated axially slotted solid rotor. For simulation a real case study is selected. It was confirmed, the speed of the HTS coated motors maintained synchronous speed up to 5 Nm constant load torque. The torque value is not achieved more than 1.7 times of the common slotted solid rotor motor. These promising properties were obtained by using of the resistance variation of the HTS material which was used as coating of slots of the solid rotor. It was concluded that the HTS coated slotted solid motor showed better performance than that of the noncoated one.

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Keywords: Bi-2223/Ag Super Conductor Critical Current Slotted Solid Rotor Starting Torque در این مقاله به بررسی موتور القایی با روتور شیاردار پوشیده شده از ابررسانای HTS پرداخته می شود. با شیاردار کردن روتور گشتاور الکترومغناطیسی در نزدیک سرعت سنکرون افزایش می یابد ولی گشتاور راهاندازی کاهش می یابد. برای بهبود گشتاور راهاندازی شیارها با روکش ابررسانا پوشیده شدهاند. این کار باعث افزایش مقاومت روتور می شود که موجب افزایش گشتاور راهاندازی خواهد شد. همچنین این پدیده موجب می شود گشتاور نزدیک سرعت سنکرون نیز افزایش یابد زیرا مقاومت روتور در این حالت تقریباً صغر می شود. با پوشش ابررسانای روتور، گشتاور راهاندازی برای موتور با روتور یکپارچه بدون شیار ۲۲ درصد و برای روتور با شیار ۷۶ درصد افزایش می یابد. همچنین گشتاور نزدیک سنکرون برای موتور با روتور یکپارچه بدون شیار ۵۷ درصد و برای روتور با شیار ۳۳ درصد برای موتور نمونه زیاد خواهد شد. عملکرد روش پیشنهادی با استفاده از روش اجزای محدود مورد شبیه سازی قرار گرفته است.

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چکیدہ