



Dimensionality Reduction and Improving the Performance of Automatic Modulation Classification using Genetic Programming

A. Latif*, K. Hessampour

Electrical and Computer Engineering Department, Yazd University, 89195741, Yazd, Iran

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ABSTRACT

This paper shows how we can make advantage of using genetic programming in selection of suitable features for automatic modulation recognition. Automatic modulation recognition is one of the essential components of modern receivers. In this regard, selection of suitable features may significantly affect the performance of the process. Simulations were conducted with 5db and 10db SNRs. Test and training data released from real ones were recorded in an actual communication system. For performance analyzing of the proposed method, a set of experiments were conducted considering signals with 2PSK, 4PSK, 2FSK, 4FSK, 16QAM and 64 QAM modulations. The results show that the selected features by the model improve the performance of automatic modulation recognition substantially. During our experiments, we also reached the suitable values and forms for mutation and crossover ratio, fitness function as well as other parameters for the proposed model.

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1. INTRODUCTION

Modulation plays an important part in electronic communication systems. In electronics and telecommunications, modulation is the process of varying one or more properties of a high frequency periodic waveform which is called carrier signal [1].

One of the newest research topics in modern communication is automatic modulation recognition. In modern receivers automatic modulation recognition is considered to be a necessary component. Recently, modulation classification is also widely used in military applications. In electronic warfare systems, this classification is used as an information source for receiving signals of enemy's communication systems [2]. In civilian application, it is mostly used in frequency management, transmitter monitoring and unauthorized transmitter detection.

Modulation recognition is an important step between signal detection and the demodulation. In non-cooperative environments, it is hardly possible to detect

the applied modulation because only little information is available in the input signal.

Automatic modulation classification is accomplished in two main steps; preprocessing of the input signal, and selection of classification method. The second step is divided into two groups. The first is maximum likelihood approach. Likelihood base classifiers are optimal in the Bayesian sense, as they minimize the probability of classification error [3]. Likelihood approaches are composite hypothesis testing problems in which classification is performed by searching for the maximum a posteriori probability $P(H_i|r(t))$, or equivalently $P(r(t)|H_i)$, given equally likely modulation schemes; that is,

$$\hat{H} = \arg \max_{H_i} \log P(r(t)|H_i) \quad (1)$$

where $r(t)$ is a function of the transmitted data, signal parameters (such as the symbol interval), the impulse response of the channel, and the noise distribution.

The latter is statistical pattern recognition approach. In this approach, first some features of the signal are extracted using mathematical methods and then the

*Corresponding Author Email: alatif@yazd.ac.ir (A. Latif)

classification would be accomplished. This approach has more diversity than the first method. Nothing is worth that generally modulation recognition methods are only applicable to some of the variants modulation. Considering above mentioned methods, the statistical ones require less computation, instead the signal features which are selected must resistant against non-ideal signal parameters [4].

Nandi and Wong have used artificial neural network and genetic algorithm for automatic modulation recognition [5]. Zhu, Aslam and Nandi have used Genetic Programming (GP) and K-nearest neighborhood for modulations classification [3]. Neshatian and Zhang have used GP for dimensionality reduction and performance improvement of standard data mining data sets [6, 7]. In this paper, it is tried to use GP intelligence selection of input features. However, this selection is done in such a way that the features number will be equal to output modulation number. Entropy concept was used for compute fitness function.

The structure of this paper is organized as follow. In section 2, signal model and signal statistical specifications have been described. The main structure of algorithm producing initial features of GP and GP structure have been introduced in section 3. In section 4, performance evaluation and result analyzing have been presented. Finally, the conclusions have been shown in section 5.

2. DATA AND MATERIAL

2. 1. Signal Model and Features For most communication systems the baseband wave form in receivers is considered as Equation (2):

$$y(n) = Ae^{j(2\pi f_0 nT + \theta_n)} \sum_{l=-\infty}^{\infty} x(l)h(nT - lT + \varepsilon_T T) + g(n) \quad (2)$$

where $x(l)$ is symbols sequence, A is amplitude, T is symbols interval, ε_T is time regulation error or sampling phase error in filter output, f_0 is deviation of receiver oscillator frequency from carrier signal, $h(\cdot)$ is baseband channel specification, θ_n is carrier signal phase deviation from receiver oscillator and $g(n)$ is additive white Gaussian noise.

In this paper, it is assumed that the working conditions is ideal and only Gaussian noise presents, the channel compensation was done, and the channel effects can be neglected.

2. 2. Features In this paper, statistical pattern recognition approach was used for modulation automatic detection. So a set of features are considered as GP search input. These features will be presented in the following section.

2. 2. 1. Key Features γ_{max} The first key feature which is shown as γ_{max} presents the maximum normal and concentrated spectrum power density for received signal instant amplitude. In accordance with Parseval's theorem, the power spectrum density of a signal equals to its DFT sequence magnitude square. It is presented as follows:

$$\gamma_{max} = \frac{\max |DFT(a_{cn}(i))|^2}{N_s} \quad (3)$$

$$a_{cn}(i) = a_n(i) - 1 \quad (4)$$

$$a_n(i) = \frac{a(i)}{m_a} \quad (5)$$

$$m_a = \frac{1}{N} \sum_{i=1}^{N_s} a(i) \quad (6)$$

N_s is the sample in each part of signal and $a_{cn}(k)$ is the normalized instant amplitude and centered related to $(t = \frac{k}{f_s}) k=1,2,\dots$, also f_s is the sampling frequency of

each part of the signal and m_a is the mean value of instant amplitude for each part of the signal.

2. 2. 2. Key Features σ_{ap} σ_{ap} is the second key feature which is equal to standard deviation of absolute number of non-linear and concentrated component of received signal instant phase and is defined as follows. In instant phase, when the received signal is higher than threshold value, the absolute standard deviation is defined as follows:

$$\sigma_{ap} = \sqrt{\frac{1}{C} \left(\sum_{a_n(i) > a_t} \phi_{NL}^2(i) \right) - \left(\frac{1}{C} \sum_{a_n(i) > a_t} |\phi_{NL}(i)| \right)^2} \quad (7)$$

where $\phi_{NL}(i)$ is the instant phase concentrated non-linear component for t so that $t = \frac{i}{f_s}$ and $i=1, 2, \dots, N_s$,

C equals to total samples related to instant phase $\phi_{NL}(i)$ which their instant amplitude is higher than the threshold value. $a_n(i)$ is the amplitude related to phase $\phi_{NL}(i)$ and a_t is the threshold value, C is the total samples which $a_n(i) > a_t$ condition is satisfied.

2. 2. 3. Key Features σ_{aa} σ_{aa} is the third key feature which is equal to standard deviation of absolute value of normalized instant amplitude when the received signal is higher than the threshold level.

$$\sigma_{aa} = \sqrt{\frac{1}{N_s} \left(\sum_{i=1}^{N_s} a_{cn}^2 \right) - \left(\frac{1}{N_s} \sum_{i=1}^{N_s} |a_{cn}| \right)^2} \quad (8)$$

2. 2. 4. Key Features σ_{af} σ_{af} is the fourth key feature which is equal to standard deviation of normalized instant frequency absolute value when the received signal is higher than the threshold level.

$$\sigma_{af} = \sqrt{\frac{1}{C} \left(\sum_{a_s(t)=a_i} f_N^2(i) \right) - \left(\frac{1}{C} \sum_{a_s(t)=a_i} |f_N(i)| \right)^2} \quad (9)$$

so that

$$f_N(i) = \frac{f(i) - m_f}{r_s} \quad (10)$$

$$m_f = \frac{1}{N_s} \sum_{i=1}^{N_s} f(i) \quad (11)$$

where r_s is the input signal symbol rate.

2. 2. 5. Second Order, Forth Order, Sixth Order and Eight Order Cumulants For a complex-valued stationary random process $y(n)$ second order, fourth order, sixth order and eighth order cumulants based on moments can be calculated according to Equation (12) [8].

3. PURPOSED METHOD

3. 1. Intelligence Features Selection In this method, it is tried to reduce the dimensionality of modulation classification using GP in order to improve the classification performance. In preprocessing step, a diagonal linear transform is applied on main features set. The steps of this operation are as follows [7]. Let us consider the input data set as D (features explained in section 2.2), $D = X \rightarrow C$ is a transform of main features set into modulation types. C is a set of different modulations $\{c_1, c_2, \dots, c_m\}$, and m is the number of different modulations, then goal is to find a new space as $D' = Y \rightarrow C$ to make the relations between the input feature space and modulation types easier. For applying class wise transformation, the training data set should be divided into m parts depending on their modulation types [7]. If each partition is considered as a spherical cloud, then its axial should be placed within modulations boundaries.

$$X_i = \{x | (x, c) \in D, c = c_i\} \text{ where } \cup_{i=1}^m \quad (13)$$

$$\begin{aligned} C_{20} &= M_{20} \\ C_{21} &= M_{21} \\ C_{40} &= M_{40} - 3M_{20}^2 \\ C_{41} &= M_{40} - 3M_{20}M_{21} \\ C_{42} &= M_{42} - |M_{20}|^2 - 2M_{21}^2 \\ C_{60} &= M_{60} - 15M_{20}M_{40} + 30M_{20}^3 \\ C_{61} &= M_{61} - 5M_{21}M_{40} - 10M_{20}M_{41} + 30M_{20}^2M_{21} \\ C_{62} &= M_{62} - 6M_{20}M_{42} - 8M_{21}M_{41} - M_{22}M_{40} + \\ & 6M_{20}^2M_{22} + 24M_{21}^2M_{20} \\ C_{63} &= M_{63} - 9M_{21}M_{42} + 12M_{21}^3 - 3M_{20}M_{43} - \\ & 3M_{22}M_{41} + 18M_{20}M_{21}M_{22} \\ C_{80} &= M_{80} - 35M_{40}^2 - 630M_{20}^4 + 420M_{20}^2M_{40} \\ C_{81} &= M_{81} - 35M_{40}M_{41} + 630M_{22}^3M_{21} + \\ & 210M_{40}M_{20}M_{21} + 210M_{20}M_{41} \\ C_{82} &= M_{82} - 15M_{40}M_{42} - 20M_{41}^2 + \\ & 30M_{40}M_{20}^2 + 60M_{40}M_{21}^2 + 240M_{41}M_{21}M_{20} \\ & + 90M_{42}M_{20}^2 - 90M_{20}^4 - 540M_{20}^2M_{21}^2 \\ C_{83} &= M_{83} - 5M_{40}M_{41} - 30M_{41}M_{42} + \\ & 90M_{41}M_{20}^2 + 120M_{41}M_{21}^2 + 180M_{42}M_{21}M_{20} \\ & + 30M_{40}M_{20}M_{21} - 270M_{20}^3M_{21} - 360M_{21}^3M_{20} \\ C_{84} &= M_{84} - M_{40}^2 - 18M_{42}^2 - 16M_{41}^2 - \\ & 54M_{20}^4 - 144M_{21}^4 - 432M_{20}^2M_{21}^2 + 12M_{40}M_{20}^2 \\ & + 96M_{41}M_{21}M_{20} - 144M_{41}M_{21}^2 + \\ & 72M_{42}M_{20}^2 + 96M_{41}M_{20}M_{21} \\ M_{pq} &= E[y(k)^{p-q} (y^*(k))^q] \end{aligned} \quad (12)$$

For finding a transform with above mentioned specifications, the data covariance should be diagonal in each part. The i^{th} covariance is computed using Equation (13):

$$\sum_i = E\{X_i X_i^T\} - E\{X_i\}E\{X_i^T\} \quad (14)$$

where \sum_i is a $n \times n$ square matrix including feature covariance depending on samples observed in i^{th} partition. This covariance matrix can be diagonal using data transform in each partition:

$$Y_i = \phi_i X, \text{ where } \phi_i = \text{eigen}(\sum_i) \quad (15)$$

This equation transforms all training dataset but the computation of covariance and eigenvector are based on the data of each partitions. If n is the input space of training dataset, then each transformation will produce n intermediate features. This process is iterated for each partition. Finally, $m \times n$ new features will be produced. Then, the D' data space will use transformed as $Y = \{Y_1, Y_2, \dots, Y_m\}$ [7]. In this step, the features of

modulation recognition and consequently the problem dimensions are increased.

This can't be used without using a suitable selection mechanism. GP search combines these elements for constructing the features of higher levels and uses a ranking function for selecting the features containing more information. The process of terminal sets construction is performed in preprocessing step. Since we have m separate modulations, $m \times n$ temporary features will be constructed and added with the main features of reference problem. They constitute the variable set of terminal pool in GP search. The variable number in intermediate buffer is $n \times (m+1)$. Dimension is increased in this step but the GP is allowed to find better feature with higher level. During the search process, an entropy-based fitness function is used to select the best features. A suitable fitness function should be able to separate informative and noninformative features. This function is used to classify all solutions in the population. Generally, entropy measurement on a modulation interval is an efficient tool for minimizing the possibility of other modulations occurrence in its interval. This measurement system has two parts: modulation limit and a metric for measuring uncertainty level in the intervals. The entropy of an interval's data is considered as a criterion for measuring the uncertainty level. Here, Shannon's entropy is used:

$$H_2 = \log_2 \sum_{c \in C} \frac{1}{p_I^2(c)} \quad (15)$$

where I is the estimated interval, C is the sum of all modulations, $P_I(c)$ is the possibility of modulation c in interval I which is computed using the occurrence frequency of modulation c in its related interval [7].

The new feature construction problem is formulated as an optimization problem for minimizing the previous equation. $\sum_{c \in C} p_I^2(c)$ should be maximized in order to

minimize the $H_2(I)$. This suggest that a feature can be considered suitable if the modulation interval only includes samples from the specific modulation. The steps for evaluating a member of population I for target modulation of t (a feature to separate the t modulation samples) on a training data set S is as follows:

A program is used to transform each sample in S data set into a transformed set S' . The main features are extracted from the terminals pool and produce a floating digit for each sample. Therefore, S' is one dimension.

An interval in S' including 99% of t modulation samples is considered as suitable interval. All samples of S' within I_t are gathered. Then $\sum_{c \in C} \frac{1}{p_I^2(c)}$ is calculated

on the data of that interval. In most samples of an interval which belong to a given modulation, the fitness function's value is low. In this design, the lower fitness

function improves the program and consequently produces better features. Only one GP run should be performed for each modulation. In each run, the fitness function is only concentrated on a given modulation. At the end of each run, the best evaluated program is selected as the feature of that modulation [7]. Therefore, the features number equals to modulation types in automatic modulation recognition.

4. RESULTS AND ANALYSIS

GPLAB tools were used in this study (<http://GPLab.sourceforge.net>). For evaluating the performance of the purposed method, the signals modulated with 10 db and 5db SNRs and 2PSK, 4PSK, 2FSK, 4FSK, 16QAM and 64 QAM modulations were studied [9]. The automatic modulation recognition involves two parts namely preprocessing and modulation classification. In the first one, the main features should be extracted. For this purpose, 600 different signals which were modulated randomly with one of the given modulations, as well as 4000 sequent samples of each signal sample which were sampled with 80 kHz sampling frequency, were gathered as training data [10]. GP was used to construct new features. The number of this feature equals to output modulations types. Six modulations were studied in this paper. Tree structure was used to present GP output. Each tree produces a floating point as the output. The GP functions set contain four basic operators $+$, $-$, \times , \div . The ramped half-and-half method was used to generate initial programs and the elitism policy was used to transfer the best parents to the next generation. The population size was 1024 and the maximum tree's depth when initializing was considered as 6. This can be increased to 12 when running. The evolution process is stopped after 25 generations. The results obtained from signal classification with main and new features (extracted from GP) were compared with each other. Multi-layer neural network perceptron was used as classifier. The 10 fold cross validation method was used for classifiers test. Best results were shown as confusion matrix in the following tables (see Table 1-8). As it can be seen, the performance of multi-layer neural network perceptron and decision tree classifier in 5db and 10 db SNRs were improved considerably using features generated by GP. This is because the GP uses entropy fitness function to find features with higher classifying ability.

5. CONCLUSIONS

This research shows that genetic programming can be used successfully in suitable feature selection for automatic modulation recognition.

TABLE 1. Modulation classification using multi-layer neural network perceptron and extracted features using GP for SNR=10db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	1	1	0	0	1	99
64QAM	3	2	0	0	98	0
2FSK	0	0	0	100	0	0
4FSK	1	0	99	0	0	0
2PSK	0	100	0	0	0	0
4PSK	100	0	0	0	0	0

TABLE 2. Modulation classification using multi-layer neural network perceptron and using main features for SNR=10db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	1	1	0	0	1	97
64QAM	3	2	0	0	95	0
2FSK	0	0	2	98	0	0
4FSK	1	0	97	1	1	0
2PSK	0	100	0	0	0	0
4PSK	99	1	0	0	0	0

TABLE 3. Modulation classification using decision tree classifier and extracted features using GP for SNR=10db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	0	1	0	0	0	99
64QAM	1	0	0	0	98	1
2FSK	0	0	0	100	0	0
4FSK	0	0	99	1	0	0
2PSK	0	100	0	0	0	0
4PSK	100	0	0	0	0	0

TABLE 4. Modulation classification using decision tree classifier and using main features for SNR=10db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	1	1	0	0	0	98
64QAM	2	0	0	0	96	2
2FSK	0	0	3	97	0	0
4FSK	0	0	96	3	0	0
2PSK	0	100	0	0	0	0
4PSK	99	1	0	0	0	0

TABLE 5. Modulation classification using multi-layer neural network and extracted features using GP for SNR=5db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	0	1	0	0	1	98
64QAM	1	2	0	0	97	0
2FSK	0	0	1	99	0	0
4FSK	1	0	98	0	0	1
2PSK	0	100	0	0	0	0
4PSK	100	0	0	0	0	0

TABLE 6. Modulation classification using multi-layer neural network perceptron and using main features for SNR=5db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	1	2	0	0	1	96
64QAM	3	2	0	0	95	0
2FSK	0	0	2	98	0	0
4FSK	1	0	97	1	1	0
2PSK	0	100	0	0	0	0
4PSK	99	1	0	0	0	0

TABLE 7. Modulation classification using decision tree classifier and extracted features using GP for SNR=5db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	0	1	0	0	1	98
64QAM	1	2	0	0	97	0
2FSK	0	0	0	100	0	0
4FSK	1	0	98	0	0	1
2PSK	0	100	0	0	0	0
4PSK	100	0	0	0	0	0

TABLE 8. Modulation classification using decision tree classifier and using main features for SNR=5db

Simulated	Extracted modulation					
	4PSK	2PSK	4FSK	2FSK	64QAM	16QAM
16QAM	1	1	0	0	1	97
64QAM	2	2	0	0	96	0
2FSK	0	0	2	98	0	0
4FSK	1	0	97	1	1	0
2PSK	0	100	0	0	0	0
4PSK	99	1	0	0	0	0

In this research, by using orthogonal class-wise transformation, we converted the input feature space into more classifiable feature space. The research also reveals that genetic programming using entropy concepts for computation of fitness function can produce informative features. Simulation results with 5db and 10db SNRs signals show that using these features, the accuracy of the multi-layer perceptron and decision tree C4.5 classifier would be improved. Also C4.5 could recognize modulation type better than the MLP classifier. In addition, the number of the produced features would be equal to the desired modulation types. This also shows that our proposed method reduces the dimension of automatic modulation recognition regardless of the input feature space.

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A. Latif, K. Hessampour

Electrical and Computer Engineering Department, Yazd University, 89195741, Yazd, Iran

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این مقاله نشان می‌دهد چگونه می‌توان با استفاده از برنامه‌نویسی ژنتیک، ویژگی‌های مناسبی را برای تشخیص خودکار نوع مدولاسیون استفاده کرد. تشخیص خودکار مدولاسیون یکی از اجزاء مهم گیرنده‌های جدید است. به همین دلیل انتخاب ویژگی‌های مناسب روی کارایی سیستم تاثیر مهمی خواهد گذاشت. شبیه‌سازی‌ها با نسبت سیگنال به نویز ۵ و ۱۰ دبی انجام شده است. داده‌های تست و آموزش از داده‌های واقعی سیستم مخابراتی واقعی تهیه شده است. برای آنالیز کارایی سیستم پیشنهادی آزمایشاتی با سیگنال‌های مدوله شده از نوع 16QAM, 4FSK, 2FSK, 4PSK, 2PSK و 64QAM انجام شده است. نتایج نشان می‌دهد که ویژگی‌های انتخاب شده توسط مدل سیستم کارایی تشخیص خودکار مدولاسیون را بهبود داده است. در حین آزمایش‌ها به مقادیر و نوع مناسبی برای نرخ تقاطع و جهش، تابع برازندگی و سایر پارامترهای مدل پیشنهادی دست یافته شد.

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